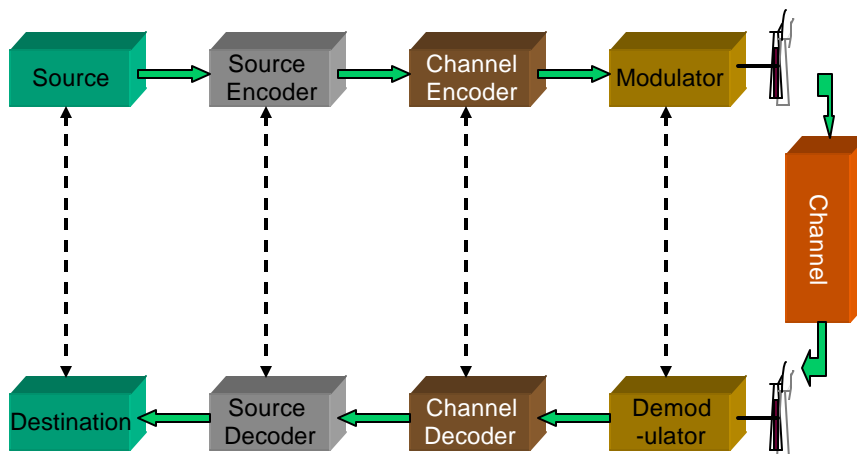


Digital Modulation

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Typical Communication System



About Channel Capacity



- Channel Capacity (C)
 - the maximum rate at which data can be transmitted over a given communication path, or channel, under given conditions
- Data rate (bps)
 - rate at which data can be communicated , impairments, such as noise, limit data rate that can be achieved
- Bandwidth (B)
 - the bandwidth of the transmitted signal as constrained by the transmitter and the nature of the transmission medium (Hertz)
- Noise (N)
 - impairments on the communications path
- Error rate - rate at which errors occur (BER)
 - Error = transmit 1 and receive 0; transmit 0 and receive 1

Reasons for Choosing Encoding Techniques



- Digital data, digital signal
 - Equipment less complex and expensive than digital-to-analog modulation equipment
- Analog data, digital signal
 - Permits use of modern digital transmission and switching equipment
- Digital data, analog signal
 - Some transmission media will only propagate analog signals
 - E.g., unguided media (air)
- Analog data, analog signal
 - Analog data in electrical form can be transmitted easily and cheaply
 - E.g., AM Radio

Signal Encoding Criteria



- What determines how successful a receiver will be in interpreting an incoming signal?
 - Signal-to-noise ratio (SNR)
 - Data rate
 - Bandwidth (B)
 - Inter-related quantities
 - Increase in SNR decreases bit error rate
 - Increase in data rate increases bit error rate
 - Increase in bandwidth allows an increase in data rate
- Shannon Bound for AWGN non fading channel

Concepts Related to Channel Capacity



- Shannon Bound for AWGN non fading channel

$$C = B \log_2 (1 + S/N)$$

- Nyquist Bandwidth
 - For binary signals (two voltage levels)
 - $C = 2B$
 - With multilevel signaling (M-ary signalling)
 - $C = 2B \log_2 M$
 - M = number of discrete signal or voltage levels
 - N = number of bits
 - $M = 2^N$

Example of Nyquist and Shannon Formulations



- Spectrum of a channel between 3 MHz and 4 MHz ; $\text{SNR}_{\text{dB}} = 24 \text{ dB}$

$$B = 4 \text{ MHz} - 3 \text{ MHz} = 1 \text{ MHz}$$

$$\text{SNR}_{\text{dB}} = 24 \text{ dB} = 10 \log_{10}(\text{SNR})$$

$$\text{SNR} = 251$$

- Using Shannon's formula

$$C = 10^6 \times \log_2(1 + 251) \approx 10^6 \times 8 = 8 \text{ Mbps}$$

- How many signaling levels are required?

$$C = 2B \log_2 M$$

$$8 \times 10^6 = 2 \times (10^6) \times \log_2 M$$

$$4 = \log_2 M$$

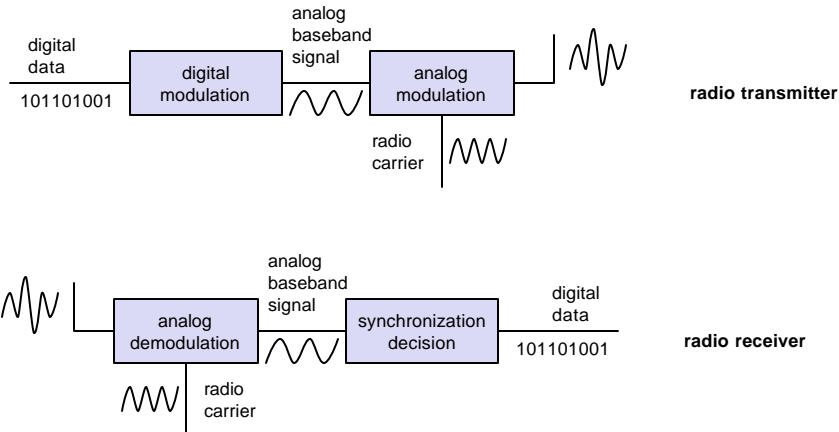
$$M = 16$$

Digital Transmission



- Why Digital ?
 - Increase System Capacity
 - compression, more efficient modulation
 - Error control coding, equalizers, etc. possible to combat noise and interference => lower power needed
 - Reduce cost and simplify designs
 - Improve Security (encryption possible)
- Digital Modulation
 - Analog signal carrying digital data

Digital Modulation and demodulation



Modulation Review



• Modulation

- Converting digital or analog information to a waveform suitable for transmission over a given medium
- Involves varying some parameter of a carrier wave (sinusoidal waveform) at a given frequency as a function of the message signal
- General sinusoid

$$A \cos(2\pi f_c t + j)$$

Amplitude Frequency Phase

- If the information is digital changing parameters is called “keying” (e.g. ASK, PSK, FSK)

Modulation

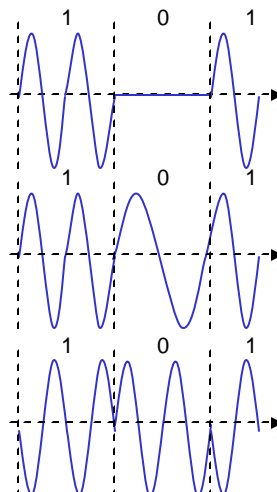


- Motivation
 - Smaller antennas (e.g., $\lambda / 4$ typical antenna size)
 - $\lambda = \text{wavelength} = c/f$, where $c = \text{speed of light}$, $f = \text{frequency}$.
 - 3000Hz baseband signal \Rightarrow 15 mile antenna, 900 MHz \Rightarrow 8 cm
 - Frequency Division Multiplexing – provides separation of signals
 - medium characteristics
 - Interference rejection
 - Simplifying circuitry
- Modulation
 - shifts center frequency of baseband signal up to the radio carrier
- Basic schemes
 - Amplitude Modulation (AM) Amplitude Shift Keying (ASK)
 - Frequency Modulation (FM) Frequency Shift Keying (FSK)
 - Phase Modulation (PM) Phase Shift Keying (PSK)

Digital modulation



- Amplitude Shift Keying (ASK):
 - change amplitude with each symbol
 - frequency constant
 - low bandwidth requirements
 - very susceptible to interference
- Frequency Shift Keying (FSK):
 - change frequency with each symbol
 - needs larger bandwidth
- Phase Shift Keying (PSK):
 - Change phase with each symbol
 - More complex
 - robust against interference



Basic Encoding Techniques

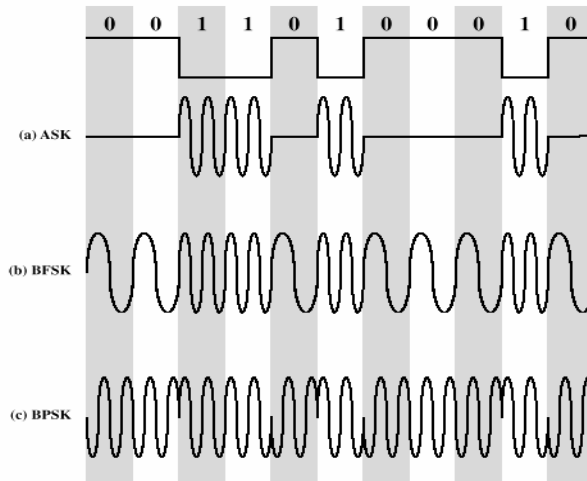


Figure 6.2 Modulation of Analog Signals for Digital Data

Amplitude-Shift Keying



- One binary digit represented by presence of carrier, at constant amplitude
- Other binary digit represented by absence of carrier

$$s(t) = \begin{cases} A\cos(2\pi f_c t) & \text{binary 1} \\ 0 & \text{binary 0} \end{cases}$$

- where the carrier signal is $A\cos(2\pi f_c t)$
- Very Susceptible to noise
- Used to transmit digital data over optical fiber

Binary Frequency-Shift Keying (BFSK)



- Two binary digits represented by two different frequencies near the carrier frequency

$$s(t) = \begin{cases} A \cos(2pf_1 t) & \text{binary 1} \\ A \cos(2pf_2 t) & \text{binary 0} \end{cases}$$

- where f_1 and f_2 are offset from carrier frequency f_c by equal but opposite amounts
- $B = 2([f_2 - f_1]/2 + f_b)$
 - Where f_b = input bit rate

Phase-Shift Keying (PSK)



- Two-level PSK (BPSK)
 - Uses two phases to represent binary digits

$$s(t) = \begin{cases} A \cos(2pf_c t) & \text{binary 1} \\ A \cos(2pf_c t + \mathbf{p}) & \text{binary 0} \end{cases}$$

$$= \begin{cases} A \cos(2pf_c t) & \text{binary 1} \\ -A \cos(2pf_c t) & \text{binary 0} \end{cases}$$

$$B = f_b$$

Selection of Encoding/Modulation Schemes



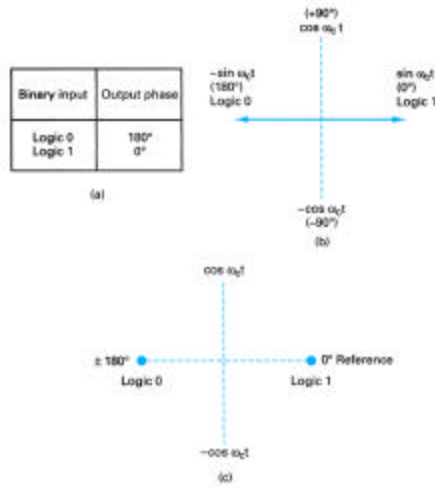
- Performance in an AWGN channel
 - How does the bit error rate vary with the energy per bit available in the system when white noise present
- Performance in fading multipath channels
 - Same as above, but add multipath and fading
- Bandwidth requirement for a given data rate
 - Also termed spectrum efficiency or bandwidth efficiency
 - How many bits/sec can you squeeze in one Hz of bandwidth for a given error rate
- Cost
 - The modulation scheme needs to be cost efficient
 - Circuitry should be simple to implement and inexpensive (e.g. detection, amplifiers)

Signal Constellation



- Given any modulation scheme, it is possible to obtain its signal constellation.
 - Represent each possible signal as a vector in a Euclidean space spanned by an orthonormal basis.
- If we know the signal constellation, we can estimate the performance in terms of the probability of symbol error or probability of bit error given the noise parameters.
- Probability of error depends on the minimum distance between the constellation points.

BPSK Signal Constellation



Tomasi
Electronic Communications Systems, 5e

Symbol Detection



- The receiver implementation can affect the performance.
 - Coherent detection
 - receiver will exploit the exact knowledge of the phase of the carrier to detect the signal better.
 - Non-coherent detection
 - involves making some approximations to the phase information that results in a loss in performance. However, it simplifies the circuitry.
- In symbol detection – decode incoming signal as closest symbol in the signal constellation space

Example of BPSK



A binary 1 is represented by:

$$s_1(t) = \sqrt{\frac{2E_b}{T}} \cos(2\mathbf{p}f_c t) \quad , \quad 0 \leq t \leq T, \quad f_c = \frac{n}{T}$$

A binary 0 is represented by:

$$s_2(t) = -\sqrt{\frac{2E_b}{T}} \cos(2\mathbf{p}f_c t) \quad , \quad 0 \leq t \leq T$$

We can write $s_1(t) = \sqrt{E_b} \Psi(t)$

$$s_2(t) = -\sqrt{E_b} \Psi(t)$$

where $\Psi(t) = \sqrt{\frac{2}{T}} \cos(2\mathbf{p}f_c t), \quad 0 \leq t \leq T$

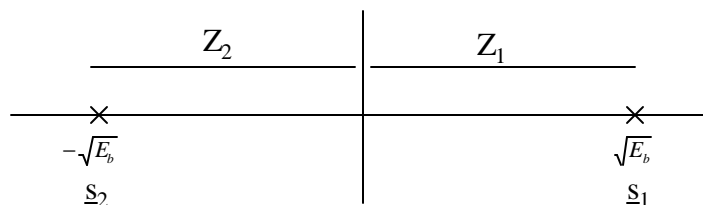
What is the energy of $\Psi(t)$?



$$\begin{aligned} E_\Psi &= \int_0^T \Psi^2(t) dt = \frac{2}{T} \int_0^T \cos^2(2\mathbf{p}f_c t) dt \\ &= \frac{2}{T} \int_0^T \frac{1}{2} [1 + \cos(4\mathbf{p}f_c t)] dt = 1 \end{aligned}$$

Note that the energy in one bit of BPSK is E_b .

The constellation of BPSK is





Detection

When do errors occur in BPSK?

- \mathbf{s}_1 is transmitted and the received point falls in \mathbf{Z}_1 region.
- \mathbf{s}_2 is transmitted and the received point falls in \mathbf{Z}_2 region.

Why will a point fall elsewhere?

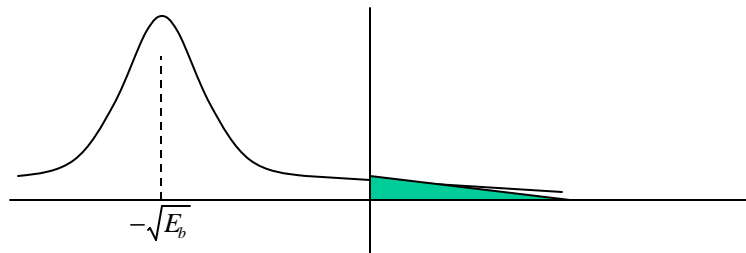
- The received signal point is $\underline{\mathbf{s}}_i + \underline{\mathbf{n}} = \underline{\mathbf{r}}$, $i=1,2$
- $\underline{\mathbf{n}}$ is the noise vector that is normally distributed with mean zero and variance $N_0/2$. This can shift the transmitted signal to some other value.



Bit Error Rate

- What is the probability that the signal point $\underline{\mathbf{r}}$ falls in \mathbf{Z}_1 given $\underline{\mathbf{s}}_2(t)$ was transmitted? (Conditional probability)
- $\underline{\mathbf{r}} = \underline{\mathbf{s}}_2 + \underline{\mathbf{n}}$ is a normally distributed random variable with mean $-\sqrt{E_b}$ and variance $N_0/2$.

$$P_e = \int_0^{\infty} \frac{1}{\sqrt{\frac{N_0}{2}} \sqrt{2p}} \exp\left[-\frac{(x + \sqrt{E_b})^2}{N_0}\right] dx$$



Normal Distribution Review



$$\text{Let } Z = \frac{(x + \sqrt{E_b})}{\sqrt{\frac{N_0}{2}}}, \quad dZ = \frac{dx}{\sqrt{\frac{N_0}{2}}}$$

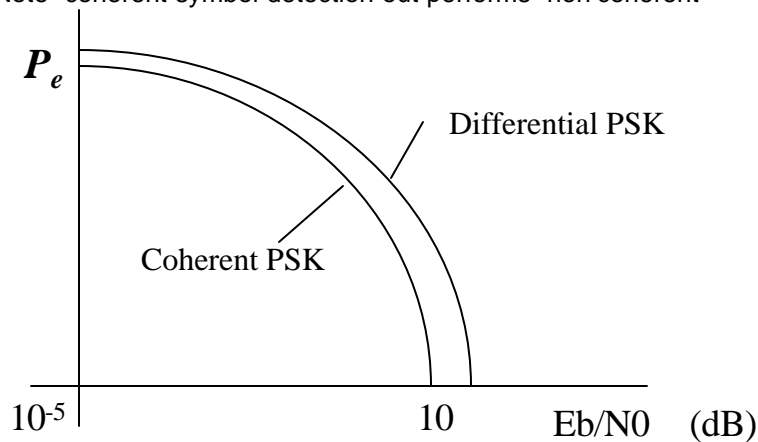
$$\text{When } x = 0, \quad Z = \sqrt{\frac{2E_b}{N_0}}$$

$$\begin{aligned} P_e &= \int_{\sqrt{\frac{2E_b}{N_0}}}^{\infty} \frac{1}{\sqrt{2\pi}} \exp\left(-\frac{Z^2}{2}\right) dz = Q\left(\sqrt{\frac{2E_b}{N_0}}\right) \\ &= Q(\mathbf{g}_b) \end{aligned}$$

PSK Error Performance



Note coherent symbol detection out performs non-coherent



Remarks

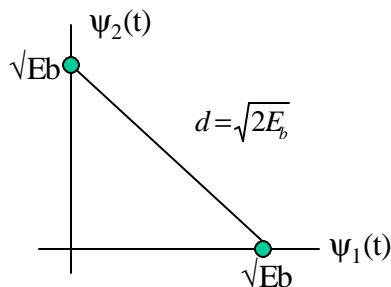


- a) We use E_b/N_0 as a figure of merit because it provides a good comparison of the “power efficiency” or “energy efficiency” of a modulation scheme. Sometimes called SNR per bit.
- b) We will not derive the bit error rate performance of different modulation schemes but we will only use the results.

Some Remarks (cont.)

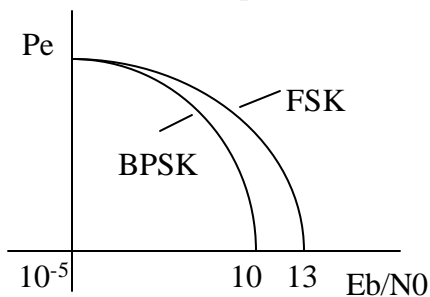


- c) The constellation of orthogonal FSK looks like this



$$P_e(FSK) = Q\left(\sqrt{\frac{E_b}{N_0}}\right)$$

(a 3dB reduction in performance)



Performance



- Bandwidth of modulated signal (B_T)

- ASK, PSK $B_T=(1+r)R$

- FSK $B_T=2DF+(1+r)R$

- R = bit rate
 - $0 < r < 1$; related to how signal is filtered
 - $DF = f_2 - f_c = f_c - f_1$

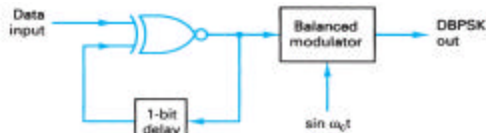
D Phase-Shift Keying (PSK)



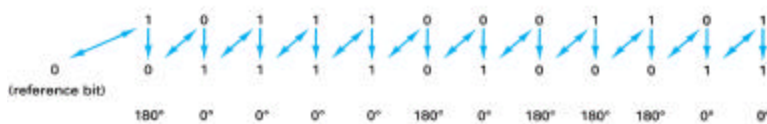
- Differential PSK (DPSK)

- Phase shift with reference to previous bit

- Binary 0 – signal burst of same phase as previous signal burst
 - Binary 1 – signal burst of opposite phase to previous signal burst



(a)

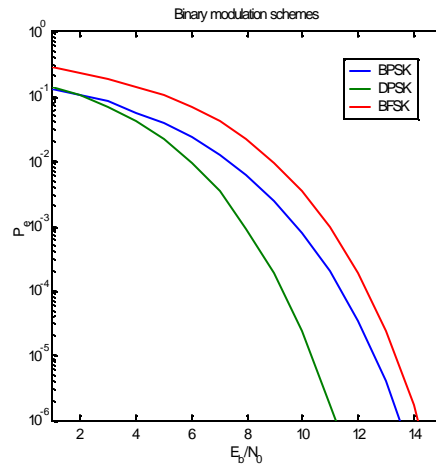


(b)

Performance in AWGN channels



- Similar to BPSK analysis have P_e for FSK, and DPSK



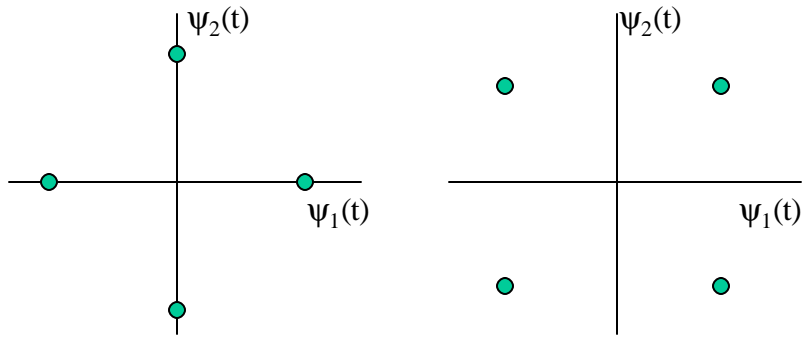
M-ary Signaling/Modulation



- What is M -ary signaling?
 - The transmitter considers 'k' bits at a times. It produces one of M signals where $M = 2^k$.
- Example: QPSK ($k = 2$)

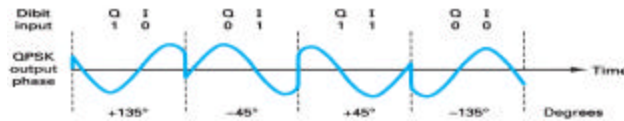
<u>Input:</u>	<u>Signal :</u>
00	$\sqrt{\frac{2E}{T}} \cos(2\mathbf{p}f_c t), \quad 0 \leq t \leq T$
01	$\sqrt{\frac{2E}{T}} \cos(2\mathbf{p}f_c t + \frac{\mathbf{p}}{2}), \quad 0 \leq t \leq T$
11	$\sqrt{\frac{2E}{T}} \cos(2\mathbf{p}f_c t + \mathbf{p}), \quad 0 \leq t \leq T$
10	$\sqrt{\frac{2E}{T}} \cos(2\mathbf{p}f_c t + \frac{3\mathbf{p}}{2}), \quad 0 \leq t \leq T$

QPSK Constellations



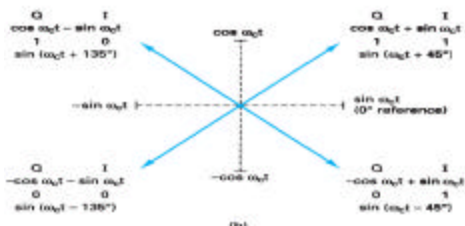
Rotated by $\pi/4$

$\pi/4$ QPSK

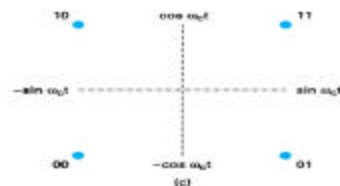


Binary input		QPSK output phase
Q	I	
0	0	-135°
0	1	-45°
1	0	$+135^\circ$
1	1	$+45^\circ$

(a)

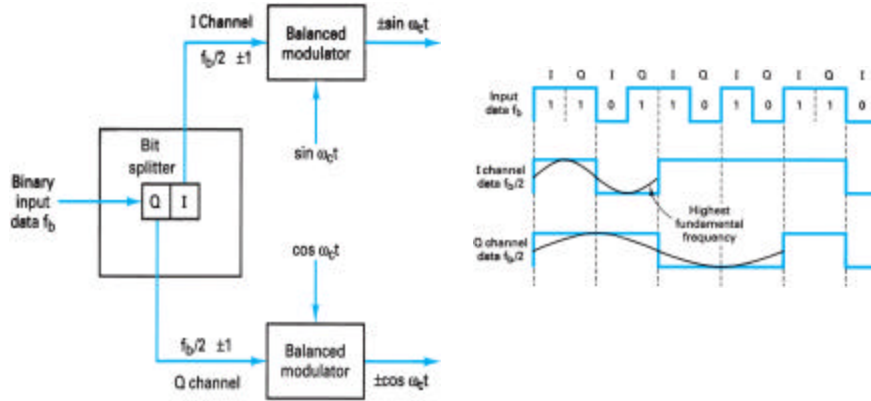


(b)



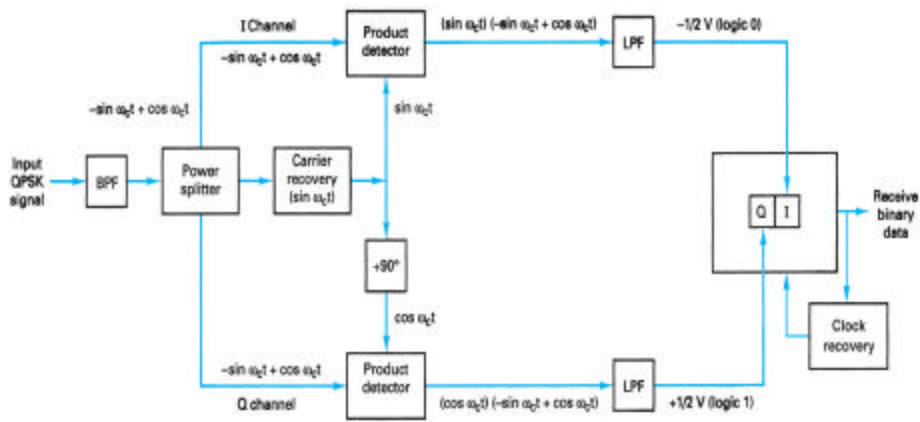
(c)

$\pi/4$ -QPSK Modulation



Can use simple AM balanced modulator

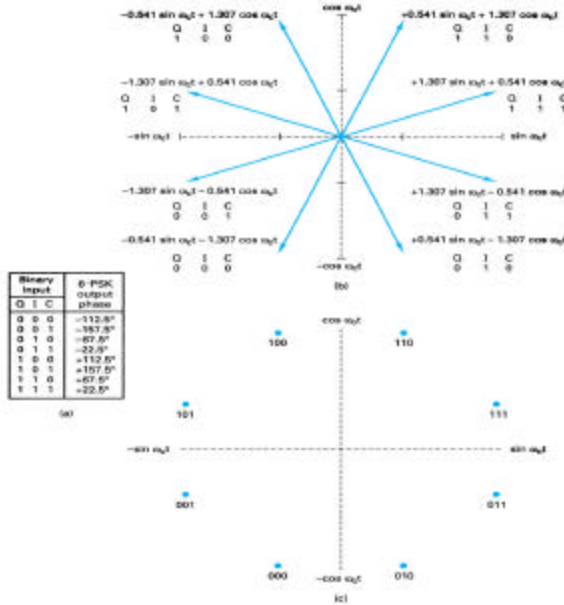
$\pi/4$ QPSK Coherent Demodulator



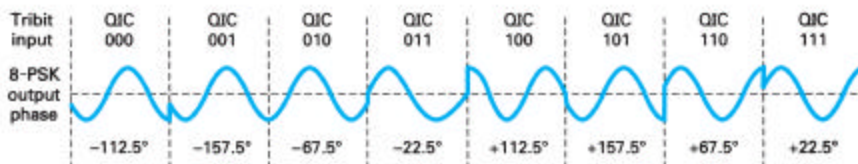


8-PSK

Increasing the number of levels increases the data rate – but increases the symbol error rate as the symbols are closer together in the constellation space



8-PSK Output



M-ary Error Performance



- MPSK, as M increases
 - the bandwidth remains constant,
 - the minimum distance between signals reduces
=> increase in symbol error rate
- MFSK, as M increases
 - the bandwidth increases
 - the performance improves but the minimum distance between signals remains the same

Performance



- Bandwidth of modulated signal (B_T)
 - MPSK
$$B_T = \left(\frac{1+r}{L} \right) R = \left(\frac{1+r}{\log_2 M} \right) R$$
 - MFSK
$$B_T = \left(\frac{(1+r)M}{\log_2 M} \right) R$$
- L = number of bits encoded per signal element
 - M = number of different signal elements

Quadrature Amplitude Modulation



- QAM is a combination of ASK and PSK
 - Two different signals sent simultaneously on the same carrier frequency –
 - Change phase and amplitude as function of input data
 - Simple case 8 QAM (two amplitudes – 4 phases)

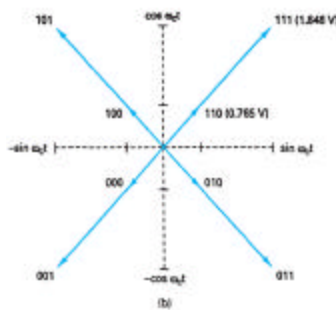
$$s(t) = d_1(t)\cos 2\pi f_c t + d_2(t)\sin 2\pi f_c t$$

8 - QAM

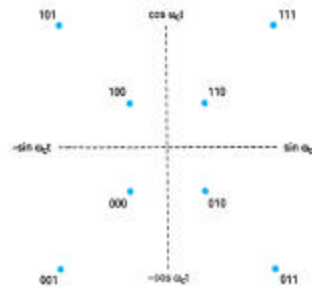


Binary input			8-QAM output	
Q	I	C	Amplitude	Phase
0	0	0	0.765 V	-135°
0	0	1	1.848 V	-135°
0	1	0	0.765 V	-45°
0	1	1	1.848 V	-45°
1	0	0	0.765 V	+135°
1	0	1	1.848 V	+135°
1	1	0	0.765 V	+45°
1	1	1	1.848 V	+45°

(a)

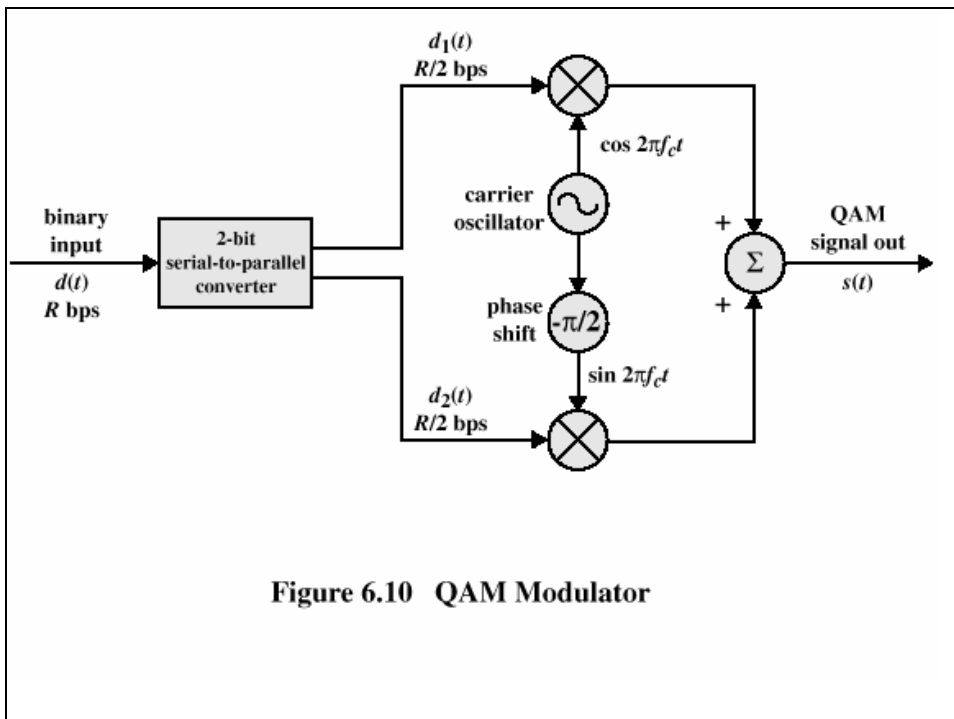
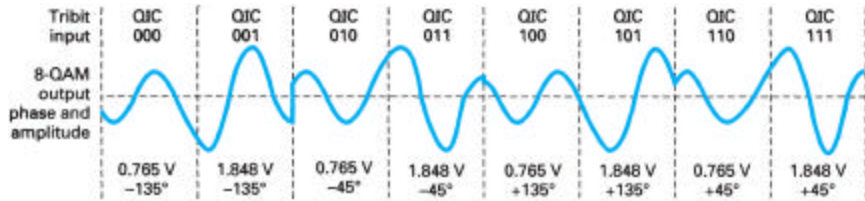


(b)



(c)

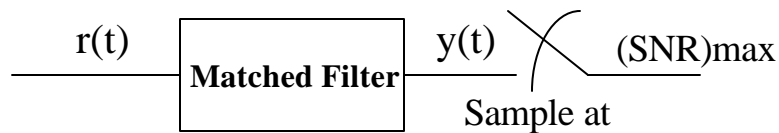
8-QAM Output



Matched Filter



- In order to detect a signal at the receiver, a linear filter that is designed to provide the maximum output SNR in AWGN for a given symbol waveform is used. This filter is called a “matched filter” (section 3.2.2)



- If the transmitted signal is $s(t)$, the impulse response of the matched filter can be shown to be

$$h(t) = \begin{cases} k \cdot s(T - t), & 0 \leq t \leq T \\ 0 & , \text{ outside} \end{cases}$$

This assumes that $s(t)$ exists only for a duration of T seconds. Let us look at the output for $k = 1$.



$$\begin{aligned} y(t) &= r(t) * h(t) \\ &= \int r(\tau) s(T - (t - \tau)) d\tau \\ &= \int r(\tau) s(T - t + \tau) d\tau \\ &= \int r(\tau) s(\tau + T - t) d\tau \end{aligned}$$

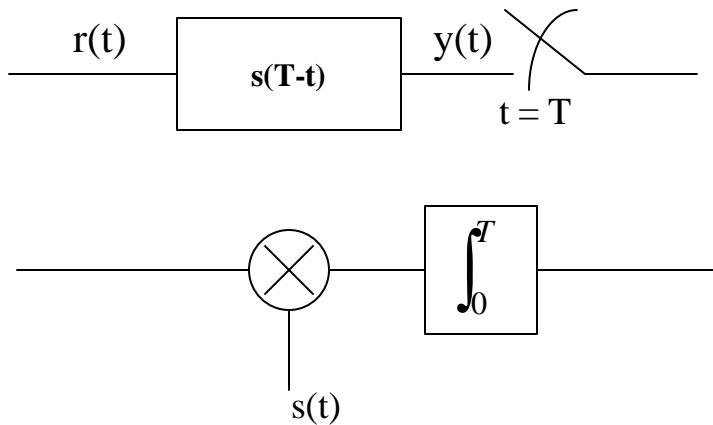
Compare with cross-correlation: $R_{rs}(t) = \int r(\tau) s(\tau - t) d\tau$

The output of the matched filter is the cross-correlation of the received signal and the time shifted transmitted signal.

$$\text{At } t = T, y(T) = \int r(\tau) s(\tau) d\tau = R_{sr}(0)$$

$$\text{If } s(t) = r(t), y(t) = E_s \text{ or } -E_s$$

Correlation Implementation of Matched Filter

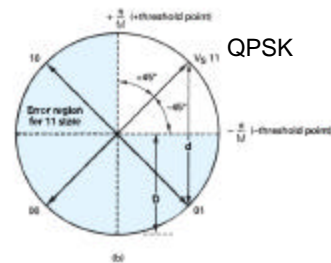
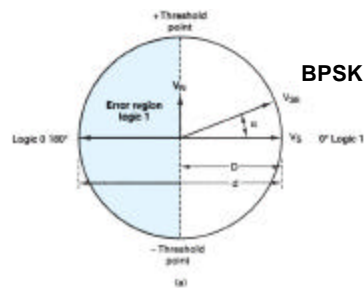


M-ary Error Performance

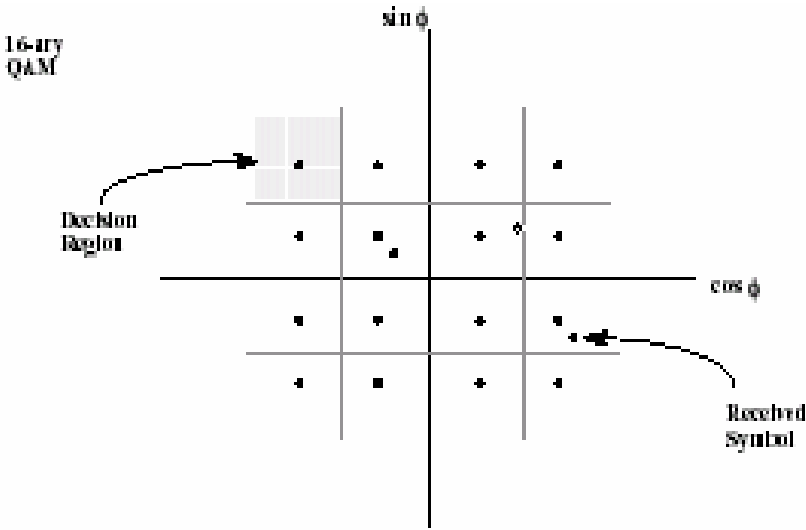


A received symbol is decoded into the closest the symbol in the signal constellation

As the number of symbols in the signal space increases the decoding region for each symbol decreases



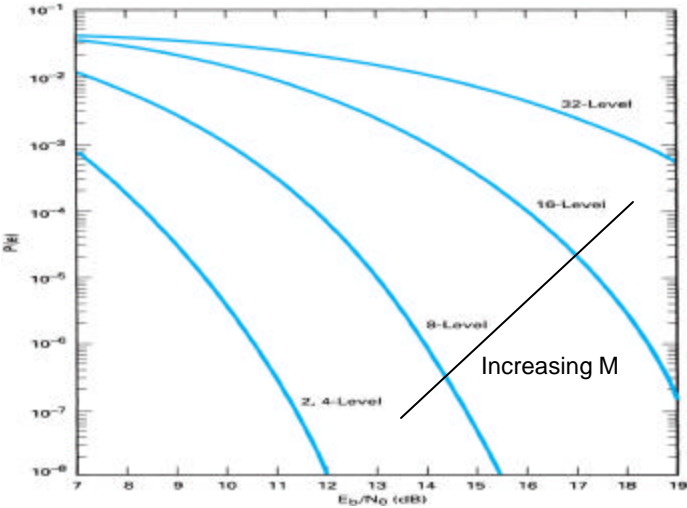
M-ary Error Performance



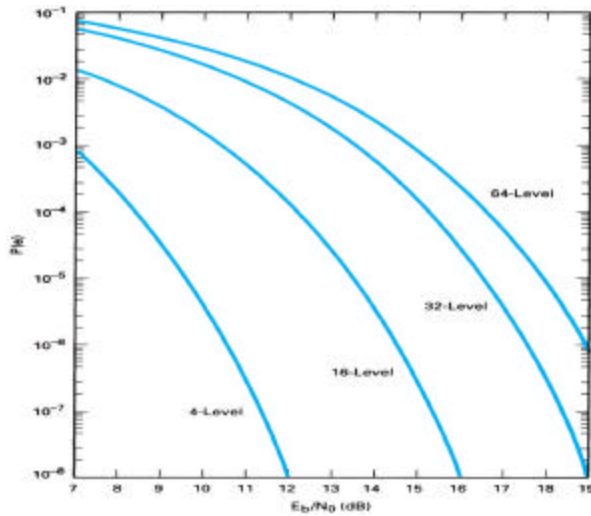
M-PSK Error Rate Performance



Increasing M increases error rate and data rate



QAM Error Performance



Performance Comparison of Various Digital Modulation Schemes (BER = 10^{-6})



Modulation Technique	C/N Ratio (dB)	E_b/N_0 Ratio (dB)
BPSK	10.6	10.6
QPSK	13.6	10.6
4-QAM	13.6	10.6
8-QAM	17.6	10.6
8-PSK	18.5	14
16-PSK	24.3	18.3
16-QAM	20.5	14.5
32-QAM	24.4	17.4
64-QAM	26.6	18.8

Tradeoffs between BER, power and bandwidth



- (1) Trade BER performance for power – fixed data rate
- (2) Trade data rate for power – fixed BER
- (3) Trade BER for data rate – fixed power

